

The Sliding Phase Vocoder

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What Moorer Said

In *Audio in the New Millennium* (JAES May 2000) James Moorer said:

Consider for one second what it means if the next 22 years gives anywhere near the same speedup as the previous 22 years has given.

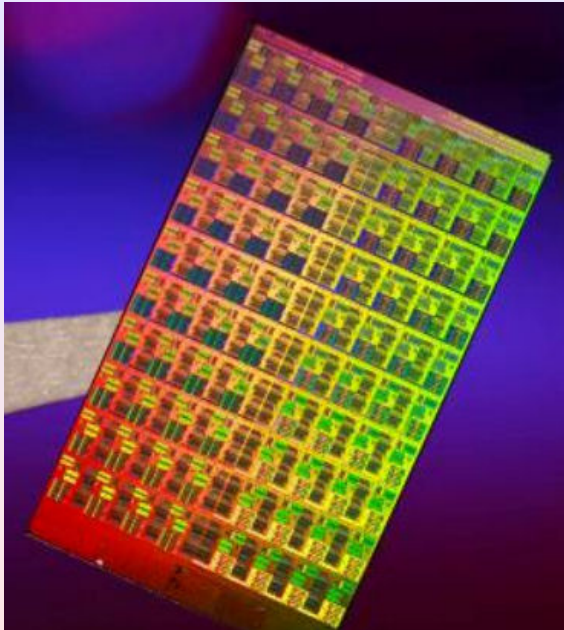
A few predictions from Moorer:

- Process 9000 channels in real time!
- Processes that were considered unfeasible...will become matter-of-fact
- Need for intelligent agents to manage 1M virtual parameters
- In 20 years, loudspeakers and microphones will know where they are.

Where are we today?

- High clock speed = high power consumption
- Multiple CPU → multiple cores — may outrun Moore's Law
- Parallel algorithms require parallel hardware
- Intel "Polaris" - 80 dual-floating point cores, 1.1TFlops on one chip ETA 10 years
- Clearspeed "Advance" - 66GFlops (64bit) @ 25 Watts, in use today
- Intel already emphasizing applications in multi-media
- Hardware expensive now, but may become commodity chip for audio acceleration
- Parallel extensions to C - new languages?

Intel Polaris 80-core



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Where are we today? (cont)

- Running transforms — the Sliding DFT
- “There is some advantage to be gained by computing the transform at every sample”
- an example ; the “ultimate sampling synthesiser”
- implication: by 2020 processing audio in the frequency domain will become routine
- compute power formidable: analysis rate = sample rate
- our interest is in interactive real-time deployment

The Sliding DFT

See for example Bradford, Dobson and Fitch, ICMC2005 Barcelona.

- Encoding: simple complex rotation - basic SDFT quite economical

$$F_{t+1}(n) = (F_t(n) - f_t + f_{t+N}) e^{2\pi i \frac{n}{N}}$$

- SDFT highly parallelizable - given the hardware
- analysis rate = sample rate so transformations inflate CPU load
- analysis: frequency-domain windowing by convolution

The Sliding DFT

- resynthesis: sum over all bins = oscillator bank

$$f_t = \frac{1}{N} \sum_{j=0}^{N-1} F_t(j) e^{-2\piijt/N}$$

but for a single point $t = 0$ this simplifies to

$$\frac{1}{N} \sum_{j=0}^{N-1} F_0(j)$$

The SPV compared to PVOOC

There are significant differences between the classic PVOOC and the Sliding Phase Vocoder (SPV):

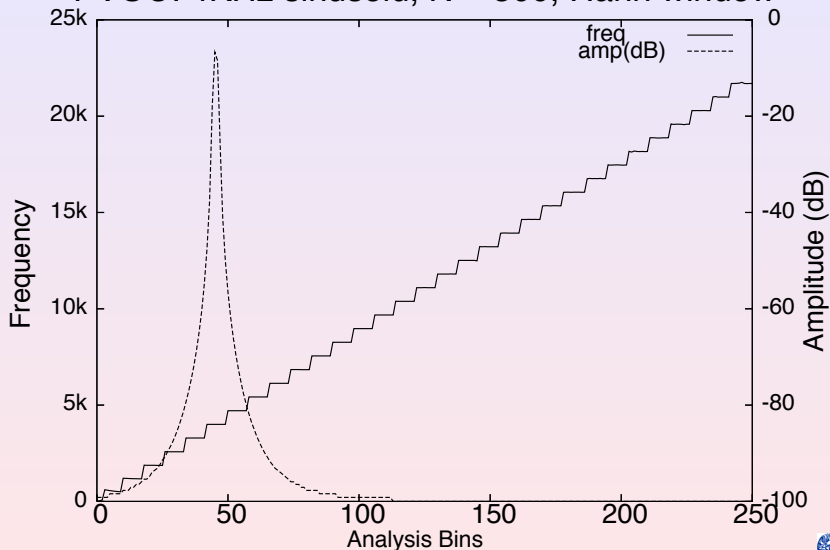
- conversion to amplitude/frequency the same
- (our SPV based on example by Stephan Sprenger (<http://www.dspdimension.org>))
- Bin bandwidth: pvoc = $\pm SR/N$; SPV = \pm Nyquist
- so in SPV: any frequency in any bin!
- pitch scaling and frequency shifting fiddly in PVOOC, trivial in SPV
- double precision floating point required as the phase updates every sample

The SPV analysis frame (amplitude/frequency)

- Basic property: frequency bunching at a peak
- pvoc — constrained by narrow bin bandwidth
- SPV: bins free to converge on peak
- extraordinary bin convergence with single input sinusoid
- bin frequency map gives new view of windowing

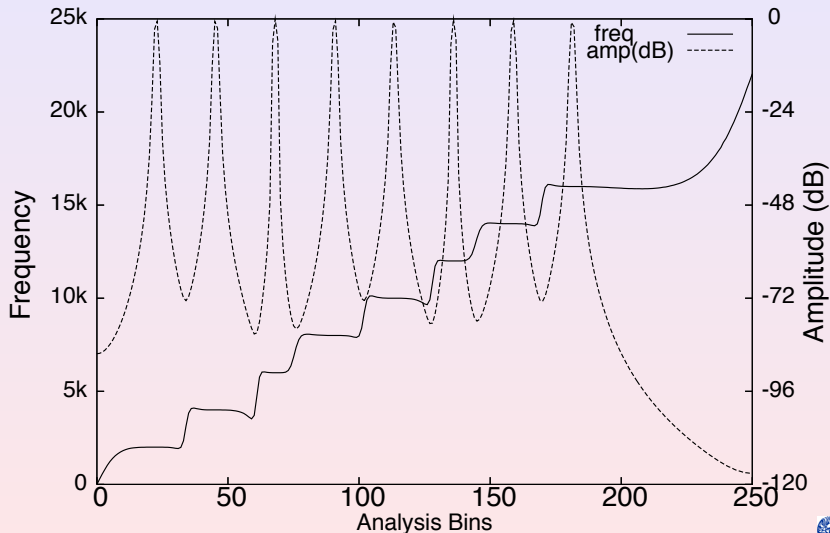
pvoc — constrained by narrow bin bandwidth

PVOC: 4KHz sinusoid, N = 500, Hann window



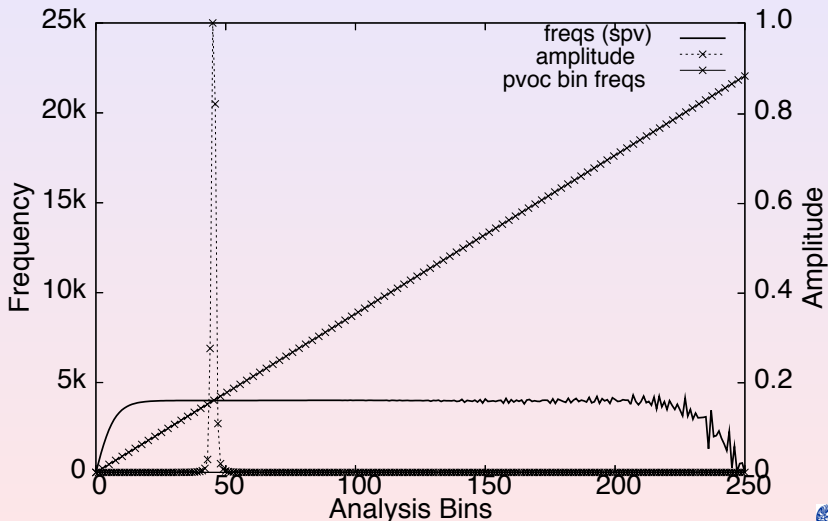
SPV: bins free to converge on peak

2KHz pulse, N = 500, Hann window

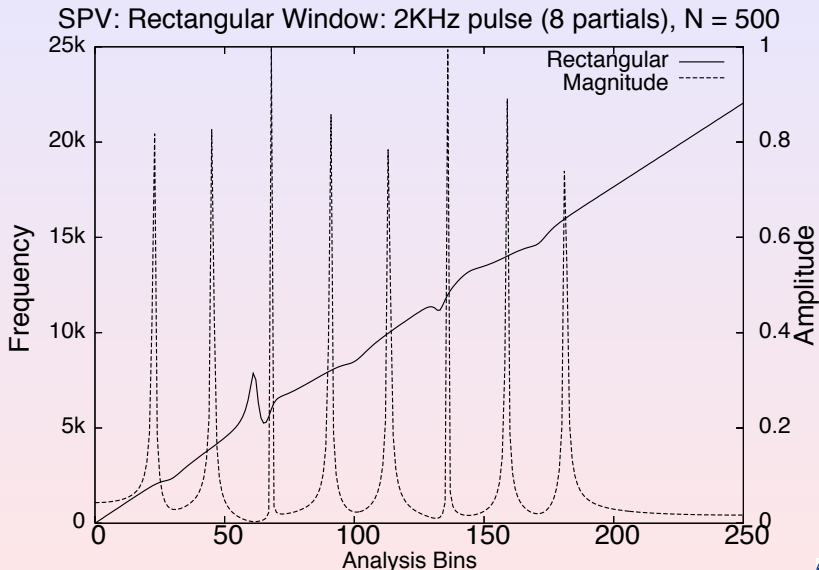


Bin convergence

4KHz sinusoid, N = 500, Hann window

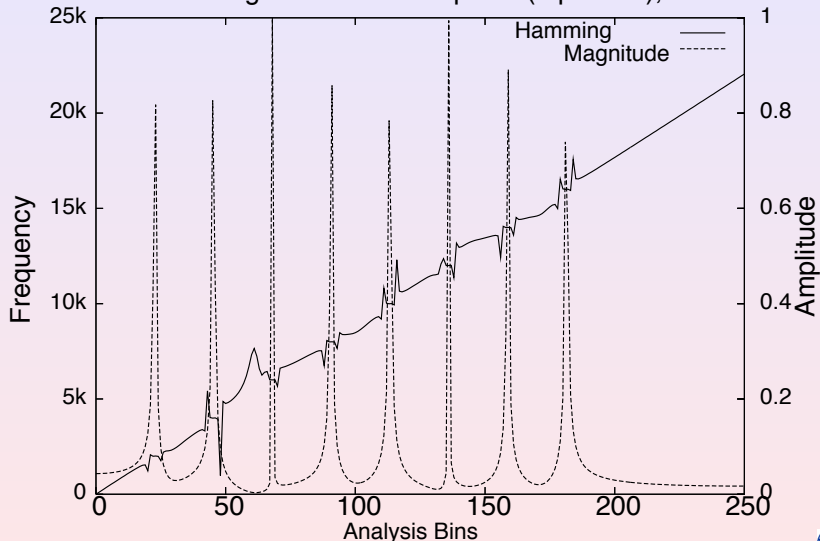


Rectangular Window



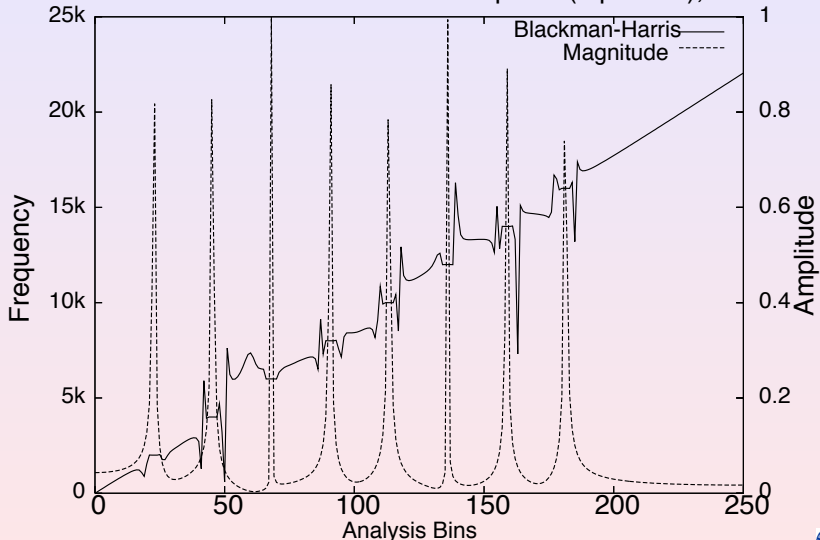
Hamming Window

SPV: Hamming Window: 2KHz pulse (8 partials), N = 500



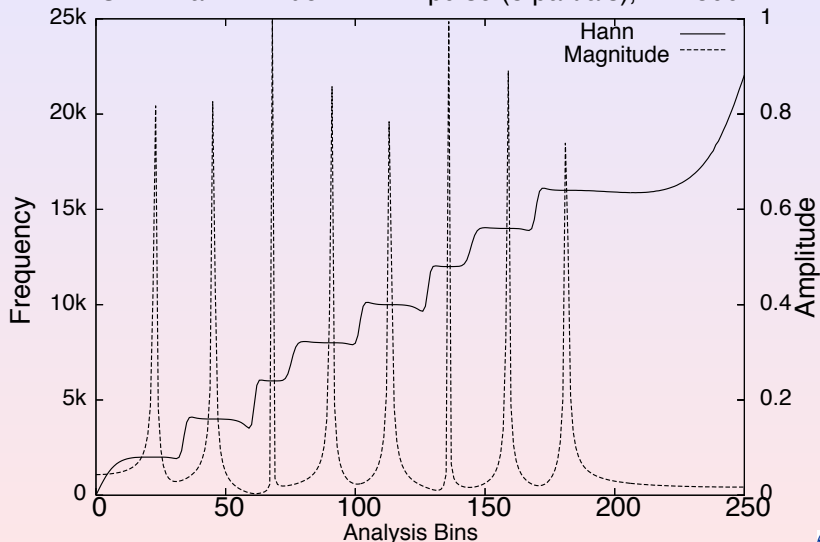
Blackman-Harris Window

SPV: Blackman-Harris Window: 2KHz pulse (8 partials), N = 500



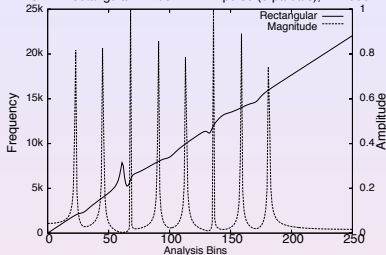
von Hann Window

SPV: Hann Window: 2KHz pulse (8 partials), N = 500

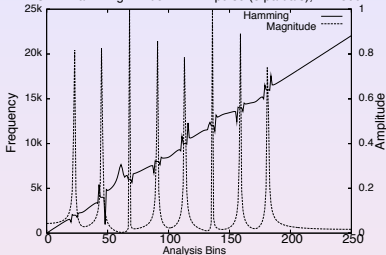


Windowing

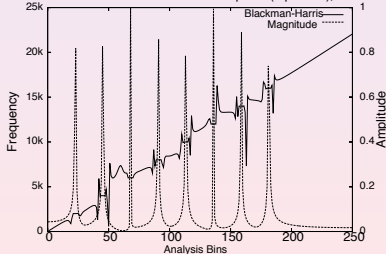
SPV: Rectangular Window: 2KHz pulse (8 partials), N = 500



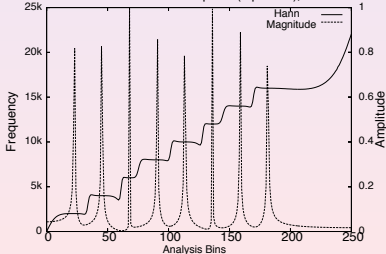
SPV: Hamming Window: 2KHz pulse (8 partials), N = 500



SPV: Blackman-Harris Window: 2KHz pulse (8 partials), N = 500



SPV: Hann Window: 2KHz pulse (8 partials), N = 500



The SPV analysis frame (amplitude/frequency)

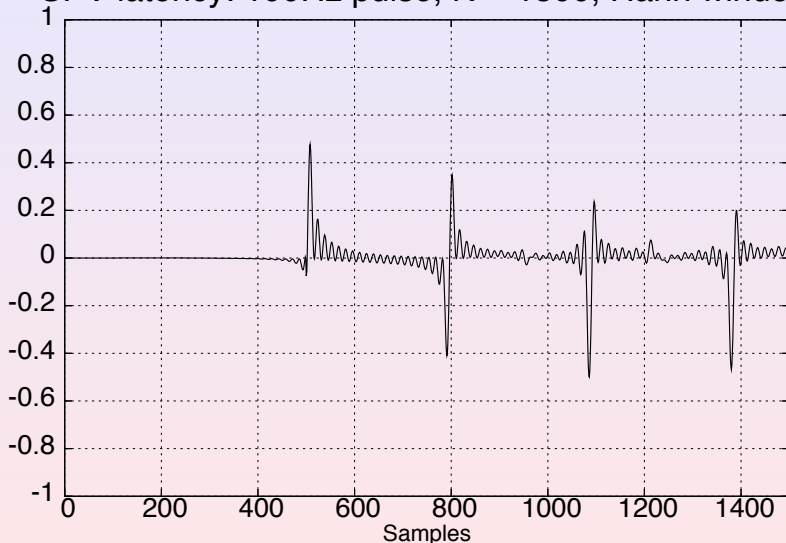
- Basic property: frequency bunching at a peak
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Advantage — what advantage?

- reduced latency - as much as $2/3^{\text{rds}}$ reduction compared to PVOC
- predictable - avoid artifacts from changing frame overlap
- frequency modifications easy, no bin-remapping required
- drums and transients - subtle differences, no free lunch
- but more sophisticated phase-locking techniques may help SPV to shine
- two ways to pitch shift:
 - simple multiplication
 - peak tracking with additive shifts

Latency

SPV latency: 100Hz pulse, $N = 1500$, Hann window



Drum Attacks

$SR = 44100$, $N = 1024$, von Hann window

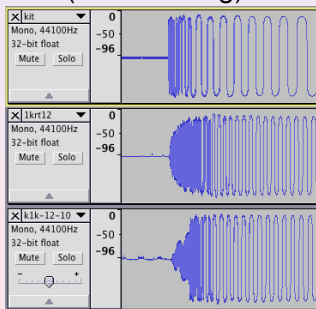
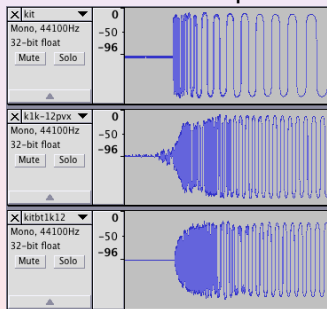
kit = source

k1k-12pvx = multiplicative (1 octave up) using standard pvoc

kitbt1k12 = multiplicative (1 octave up) using SPV

1krt12 = SPV tracking shift one octave

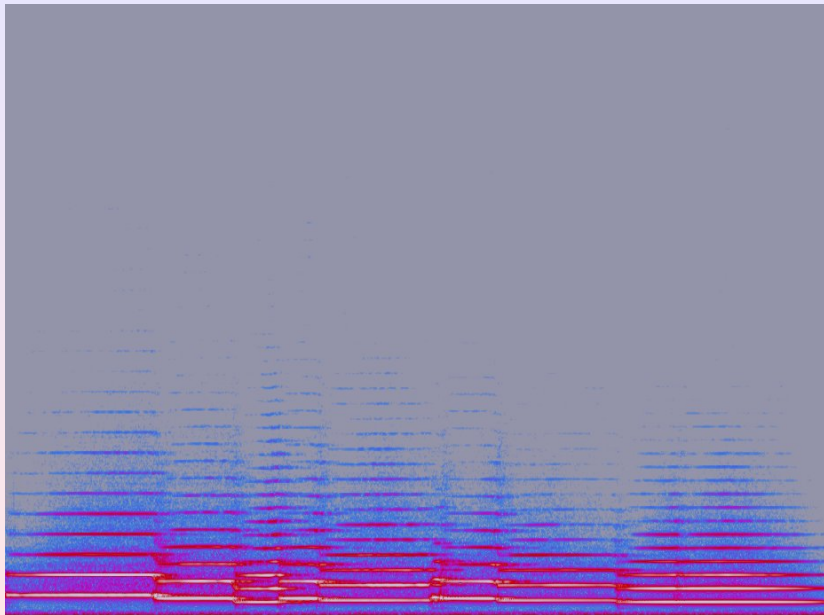
k1k-12-10 = SPV plain additive (non-tracking) shift by 100 Hz



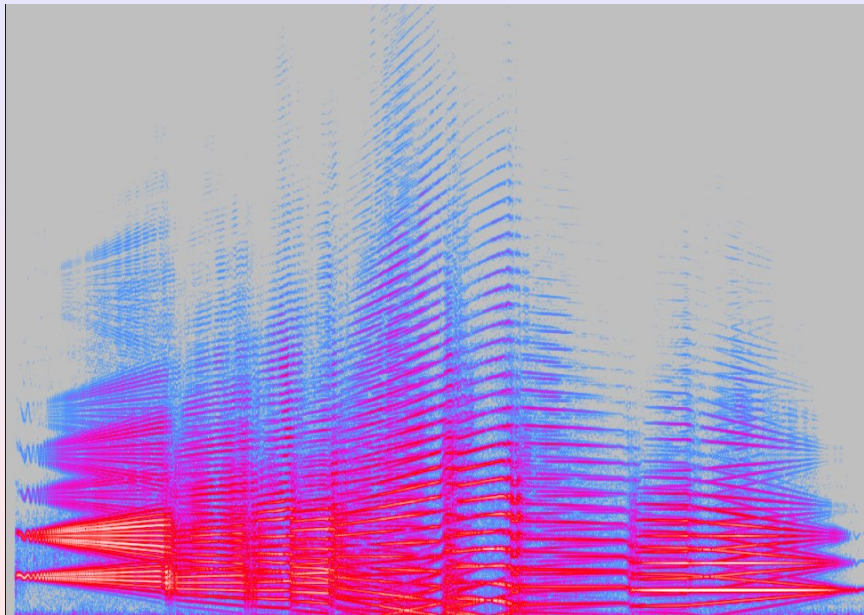
Advantage — what advantage? (cont)

- two ways to pitch shift:
 - simple multiplication
 - peak tracking with additive shifts
- simple additive shift very effective with drums (*sound example*)
- anything you can do with an oscillator bank, you can do with the SPV
- single-sample update enables transformational FM (*sound example*)

Horn Spectrum



Horn Spectrum with FM



Our current implementation is on stock Linux and Macintosh computers.

- examples on standard hardware slow!
- expect Advance cards to enable real-time use
- will place code on the net



We are advocating the study of High Performance Audio Computing (HiPAC)

- focus on algorithms that are highly parallelizable (SIMD model)
- SPV is example of “no-compromise” process
- still life in PVOOC - multiple channels?
- SPV special case of additive synthesis
- e.g. direct computation of b/l pulse, square, triangle waves
- Holy Grail = constant-Q (with inverse)
- other processes:
 - physical models based on meshes, finite differences
 - instruments, spaces
 - chance to explore no-compromise “ideal” algorithms

We finish with a question:

What would you do with 1TFlops?

Acknowledgment

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